A Reconstruction Algorithm for Photoacoustic Imaging based on the Nonuniform FFT

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Abstract

Fourier reconstruction algorithms significantly outperform conventional back-projection algorithms in terms of computation time. In photoacoustic imaging, these methods require interpolation in the Fourier space domain, which creates artifacts in reconstructed images. We propose a novel reconstruction algorithm that applies the one-dimensional nonuniform fast Fourier transform to photoacoustic imaging. It is shown theoretically and numerically that our algorithm avoids artifacts while preserving the computational effectiveness of Fourier reconstruction.

Key-words. Image reconstruction, photoacoustic imaging, planar measurement geometry, fast algorithm, nonuniform FFT.

1 Introduction

Photoacoustic imaging (PAI) is a novel promising tool for visualizing light absorbing structures in an optically scattering medium, which carry valuable information for medical diagnostics. It is based on the generation of acoustic waves by illuminating an object with pulses of non-ionizing electromagnetic radiation, and combines the high contrast of pure optical and the high resolution of ultrasonic imaging. The method has demonstrated great promise for a variety of biomedical applications, such as imaging of animals [1, 2], early cancer diagnostics [3, 4], and imaging of vasculature [5, 6].

When an object is illuminated with short pulses of non-ionizing electromagnetic radiation, it absorbs a fraction of energy and heats up. This in turn induces acoustic (pressure) waves, that are recorded with acoustic detectors outside of the object. Other than in conventional ultrasound imaging, where the source of acoustic waves is an external transducer, in PAI the source is the imaged object

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itself. The frequency bandwidth of the recorded signals is therefore generally broad and depends on the size and the shape of illuminated structures.

1.1 Planar recording geometry

Throughout this paper we assume a planar recording geometry, where the acoustic signals are recorded with omnidirectional detectors arranged on planes (or lines), see Figure 1. The planar geometry is of particular interest since it can be realized most easily in practical applications. The recorded acoustic signals are then used to reconstruct the initially generated acoustic pressure which represents optically absorbing structures of the investigated object.

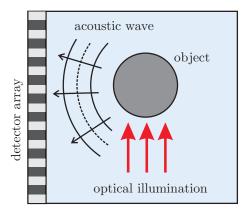


Figure 1: Photoacoustic imaging for planar recording geometry. The object is illuminated by a pulse of electromagnetic radiation, and reacts with an expansion. Induced acoustic waves are measured with an array of acoustic detectors arranged on a plane (or a line) and used to from an image of the object.

For the planar recording geometry, two types of theoretically exact reconstruction formulas have been reported: Temporal back-projection [7, 8, 9, 10] and Fourier domain formulas [11, 7, 9, 12, 13, 14, 15]. Numerical implementations of those formulas often lead to fast and accurate image reconstruction algorithms.

In temporal back-projection formulas the signals measured at time t are back projected over spheres of radius $v_s t$ with the detector position in the centre (v_s denotes the speed of sound). In Fourier domain formulas this back-projection is performed by interpolation in the frequency domain. Reconstruction methods based on Fourier domain formulas are attractive since they reconstruct an $N \times N \times N$ image in $\mathcal{O}(N^3 \log N)$ floating point operations by using of the Fast Fourier Transform (FFT). Straightforward implementations of back-projection type formulas, on the other hand, require $\mathcal{O}(N^5)$ operation counts, see [16, 17].

The standard FFT algorithm assumes sampling on an equally spaced grid and therefore, in order to implement the Fourier domain formulas, interpolation

in the Fourier space is required. Interpolation in the Fourier domain is a critical issue, and creates artifacts in reconstructed images, see the examples in Section 5. One obtains significantly better results by increasing the sampling density in the Fourier space. This is achieved by either zero-padding [13] or by symmetrizing the recorded signals around t=0 (which is equivalent to using the fast cosine transform instead of the FFT). In this paper, we propose an efficient reconstruction algorithm that uses the nonuniform (or unevenly spaced) FFT [18, 19, 20, 21, 22, 23] and further increases the quality of reconstruction.

1.2 Prior work and innovations

The nonuniform FFT has been applied to a variety of medical imaging problems, such as standard X-ray CT, magnetic resonance imaging, and diffraction tomography [24, 25, 26], and has also been used implicitly in gridding algorithms [27, 28]. All those algorithms deal with the problem of recovering a two (or higher) dimensional object function from samples of its multidimensional Fourier transform on a non-cartesian grid.

Our approach is conceptually quite different to the above mentioned references: The special structure of our problem allows to perform several one-dimensional nonuniform FFTs instead of a single higher dimensional one. This leads to a reduced numerical cost, compared to the above algorithms. The proposed algorithm is more closely related to a reconstruction algorithm for X-ray CT suggested in [29, Section 5.2], which also evaluates the Fourier transform on irregular samples by means of the one-dimensional FFT.

1.3 Outline

This article is organized as follows: In Section 2 we present the mathematical basics of Fourier reconstruction in PAI. In Section 3 we review the nonuniform FFT which is then used to derive the nonuniform FFT based reconstruction algorithm in Section 4. In Section 5 we present numerical results of the proposed algorithm and compare it with existing Fourier and back projection algorithms. The paper concludes with a discussion of some issues related to sampling and resolution in the Appendix.

2 Photoacoustic Imaging

Let $C_0^{\infty}(\mathbb{H})$ denote the space of smooth functions with bounded support in the half space $\mathbb{H} := \mathbb{R}^{d-1} \times (0, \infty), d \geq 2$. Consider the initial value problem

$$(\partial_t^2 - \Delta) p(\mathbf{x}, t) = 0, \qquad (\mathbf{x}, t) \in \mathbb{R}^d \times (0, \infty),$$
$$p(\mathbf{x}, 0) = f(\mathbf{x}), \qquad \mathbf{x} \in \mathbb{R}^d,$$
$$\partial_t p(\mathbf{x}, 0) = 0, \qquad \mathbf{x} \in \mathbb{R}^d,$$

with $f \in C_0^{\infty}(\mathbb{H})$. Here Δ denotes the Laplacian in \mathbb{R}^d and ∂_t is the derivative with respect to t. We write $\mathbf{x} = (x, y), x \in \mathbb{R}^{d-1}, y \in \mathbb{R}$, and define the operator $\mathbf{Q}: C_0^{\infty}(\mathbb{H}) \to C^{\infty}(\mathbb{R}^d)$ by

$$(\mathbf{Q} f)(x,t) := \left\{ \begin{array}{ll} p(x,y=0,t) \,, & \text{if } t > 0 \,, \\ 0 \,, & \text{otherwise} \,. \end{array} \right.$$

Photoacoustic imaging for planar recording geometry is concerned with reconstructing $f \in C_0^{\infty}(\mathbb{H})$ from incomplete and possibly erroneous knowledge of $\mathbf{Q} f$. Of practical interest are the cases d=2 and d=3, see [30, 31, 32, 33].

2.1 Exact inversion formula

The operator \mathbf{Q} can be inverted analytically by means of the exact inversion formula

$$(\mathbf{F} f)(K_x, K_y) = \frac{2K_y(\mathbf{F} \mathbf{Q} f) \left(K_x, \operatorname{sign}(K_y) \sqrt{|K_x|^2 + K_y^2}\right)}{\operatorname{sign}(K_y) \sqrt{|K_x|^2 + K_y^2}}$$
(1)

where $(K_x, K_y) \in \mathbb{R}^{d-1} \times \mathbb{R}$, and **F** denotes the d-dimensional Fourier transform,

$$(\mathbf{F}\,\varphi)(\mathbf{K}) := \int_{\mathbb{R}^d} e^{-i\mathbf{K}\mathbf{x}} \varphi(\mathbf{x}) \, d\mathbf{x}, \qquad \mathbf{K} = (K_x, K_y) \in \mathbb{R}^d.$$

Equation (1) has been derived in [12, 13] for three spatial dimensions. It can be proven in any dimension by using the inversion formula for the spherical mean Radon transform of [7, 9]. A related formula using the Fourier cosine transform instead of the Fourier transform has been obtained in [34, 15] for d = 2, 3.

2.2 Partial Fourier reconstruction

The inversion formula (1) yields an exact reconstruction of f, provided that $(\mathbf{Q} f)(x,t)$ is given for all $(x,t) \in \mathbb{R}^d$. In practical applications only a partial (or limited view) data set is available [35, 36, 37]. In this paper we assume that data $(\mathbf{Q} f)(x,t)$ are given only for $(x,t) \in (0,X)^d$, see Figure 1, which are modeled by

$$q(x,t) := w_{\text{cut}}(x,t)(\mathbf{Q} f)(x,t), \qquad (2)$$

where w_{cut} is a smooth nonnegative cutoff function that vanishes outside $(0, X)^d$. Using data (2), the function f cannot be exactly reconstructed in a stable way (see [38, 37]). It is therefore common to apply the exact inverse of \mathbf{Q} to the partial data g and to consider the result as an approximation of the object to be reconstructed. More precisely, the function f^{\dagger} defined by

$$(\mathbf{F} f^{\dagger})(K_x, K_y) := \frac{2K_y(\mathbf{F} g) \left(K_x, \operatorname{sign}(K_y) \sqrt{|K_x|^2 + K_y^2}\right)}{\operatorname{sign}(K_y) \sqrt{|K_x|^2 + K_y^2}},$$
(3)

is considered an approximation of f. The function f^{\dagger} is called *partial Fourier reconstruction*.

Fourier reconstruction algorithms in PAI name numerical implementations of (3). In this paper we apply the one-dimensional nonuniform FFT to derive a fast and accurate algorithm for implementing (3).

3 The Nonuniform Fast Fourier Transform

The discrete Fourier transform of a vector $\mathbf{g} = (g_n)_{n=0}^{N-1} \in \mathbb{C}^N$ with respect to the nodes $\boldsymbol{\omega} = (\omega_k)_{k=-N/2}^{N/2-1}$ (with N even) is defined by

$$T[\mathbf{g}](\omega_k) := \sum_{n=0}^{N-1} e^{-i\omega_k n 2\pi/N} g_n , \quad k = -N/2, \dots, N/2 - 1.$$
 (4)

Direct evaluation of the N sums in (4) requires $\mathcal{O}(N^2)$ operations. Using the classical fast Fourier transform (FFT) this effort can be reduced to $\mathcal{O}(N \log N)$ operations. However, application of the classical FFT is restricted to the case of evenly spaced nodes $\omega_k = k, k = -N/2, \dots, N/2 - 1$.

The one-dimensional nonuniform FFT (see [18, 19, 20, 29, 21, 22, 23]) is an approximate but highly accurate method for evaluating (4) at arbitrary nodes ω_k , $k = -N/2, \ldots, N/2 - 1$, in $\mathcal{O}(N \log N)$ operations.

3.1 Derivation of the nonuniform FFT

To derive the nonuniform FFT we closely follow the presentation of [29], which is based on the following lemma:

Lemma 1 ([29, Proposition 1]). Let c > 1 and $\alpha < \pi(2c - 1)$. Assume that $\Psi : \mathbb{R} \to \mathbb{R}$ is continuous in $[-\alpha, \alpha]$, vanishing outside $[-\alpha, \alpha]$, and positive in $[-\pi, \pi]$. Then

$$e^{-i\omega\theta} = \frac{c}{2\pi\Psi(\theta)} \sum_{j\in\mathbb{Z}} \hat{\Psi}(\omega - j/c) e^{-ij\theta/c}, \quad \omega \in \mathbb{R}, |\theta| \le \pi.$$
 (5)

Here $\hat{\Psi}(\omega) := \int_{\mathbb{R}} e^{-i\omega\theta} \Psi(\theta) d\theta$ denotes the one-dimensional Fourier transform of Ψ .

Proposition 2. Let c, α , Ψ , and $\hat{\Psi}$ be as in Lemma 1. Then, for every $\mathbf{g} = (g_n)_{n=0}^{N-1} \in \mathbb{C}^N$ and $\omega \in \mathbb{R}$ we have

$$\sum_{n=0}^{N-1} e^{-i\omega n 2\pi/N} g_n = \sum_{j \in \mathbb{Z}} e^{-i\pi(\omega - j/c)} \hat{\Psi}(\omega - j/c) \hat{G}_j, \qquad (6)$$

with

$$\hat{G}_j := \frac{c}{2\pi} \left(\sum_{n=0}^{N-1} \frac{g_n e^{-ijn2\pi/(Nc)}}{\Psi(n2\pi/N - \pi)} \right), \qquad j \in \mathbb{Z}.$$
 (7)

Proof. Taking $\theta = n2\pi/N - \pi \in [-\pi, \pi]$ in (5), gives

$$e^{-i\omega n2\pi/N} = \frac{c}{2\pi\Psi(n2\pi/N-\pi)} \sum_{j\in\mathbb{Z}} \hat{\Psi}(\omega-j/c) e^{-ijn2\pi/(cN)} e^{-i\pi(\omega-j/c)} ,$$

and therefore

$$\sum_{n=0}^{N-1} e^{-i\omega n 2\pi/N} g_n = \frac{c}{2\pi} \sum_{n=0}^{N-1} \sum_{j \in \mathbb{Z}} e^{-i\pi(\omega - j/c)} \hat{\Psi}(\omega - j/c) \frac{g_n e^{-ijn2\pi/(cN)}}{\Psi(n2\pi/N - \pi)}.$$

Interchanging the order of summation in the right hand side of the above equation shows (6), (7) and concludes the proof.

In the following we assume that cN is an even number. Then

$$\hat{G}_{j} = \frac{c}{2\pi} \left(\sum_{n=0}^{cN-1} \frac{g_{n}}{\Psi(n2\pi/N - \pi)} e^{-ijn2\pi/(Nc)} \right), \quad j \in \mathbb{Z},$$
 (8)

where $g_n := 0$ for $n \ge N$, is an oversampled discrete Fourier transform with the oversampling factor c. Moreover we assume that $\hat{\Psi}$ is concentrated around zero and decays rapidly away from zero. The nonuniform FFT uses the formulas (6), (8) to evaluate $T[\mathbf{g}]$ at the nodes ω_k . The basic steps of the algorithm are as follows:

- (i) Append (c-1)N zeros to the vector $\mathbf{g} = (g_n)_{n=0}^{N-1}$ and evaluate \hat{G}_j , $j = -Nc/2, \ldots, Nc/2-1$, in (8) with the FFT algorithm.
- (ii) Evaluate the sums in (6) approximately by using only the terms with $|\omega_k j/c| \leq K$, where the *interpolation length* K is a small positive parameter.

Since $\hat{\Psi}$ is assumed to decay rapidly, the truncation error in Step (ii) is small.

The nonuniform Fourier transform is summarized in Algorithm 1. All evaluations of Ψ and $\hat{\Psi}$ are precomputed and stored. Moreover, the classical FFT is applied to a vector of length cN. Therefore the numerical complexity of Algorithm 1 is $\mathcal{O}(cN\log N)$. Typically c=2, in which case the numerical effort of the nonuniform FFT is essentially twice the effort of the one-dimensional classical FFT applied to an input vector of the same length. See [29, Section 3] for an exact operation count, and a comparison between actual computation times of the classical and the nonuniform FFT.

3.2 Kaiser Bessel window

In our implementation we choose for Ψ the Kaiser Bessel window,

$$\Psi_{\mathrm{KB}}^{\alpha,K}(\theta) := \frac{1}{I_0(\alpha K)} \left\{ \begin{array}{l} I_0(K\sqrt{\alpha^2 - \theta^2}) \,, & \text{if } |\theta| \leq \alpha \,, \\ 0 \,, & \text{if } |\theta| > \alpha \,. \end{array} \right.$$

Algorithm 1 Nonuniform FFT with respect to the nodes $\boldsymbol{\omega} = (\omega_k)_{k=-N/2}^{N/2-1}$, using input vector $\mathbf{g} = (g_n)_{n=0}^{N-1}$, oversampling c > 1, interpolation length K, and window function Ψ .

```
1: \Psi \leftarrow (\Psi(2\pi n/N - \pi))_n
                                                                                                                                   ▷ precomputations
 2: \hat{\boldsymbol{\Psi}} \leftarrow \left(e^{-i(\omega_k - j/c)\pi/c}\hat{\boldsymbol{\Psi}}(\omega_k - j/c)\right)_k
  4: function nufft(\mathbf{g}, \boldsymbol{\omega}, c, K, \boldsymbol{\Psi}, \hat{\boldsymbol{\Psi}})
               \mathbf{g} \leftarrow \mathbf{g}/\mathbf{\Psi} \cdot c/(2\pi)
               \mathbf{g} \leftarrow (\mathbf{g}, \mathtt{zeros}(1, (c-1)N))
                                                                                                                                           ▷ zero-padding
  6:
               \mathbf{g} \leftarrow \mathtt{fft}(\mathbf{g})
                                                                                                                         ▷ one-dimensional FFT
  7:
              for k = -N/2, \ldots, N/2 - 1 do \hat{g}_k \leftarrow \sum_{|j-c\omega_k| \le cK} \hat{\Psi}_{k,j} g_j end for
  8:
                                                                                                                                           ▶ interpolation
  9:
10:
               return (\hat{g}_k)_k
12: end function
```

Here I_0 is the modified Bessel function of order zero. The one-dimensional Fourier transform of $\Psi_{\rm KB}^{\alpha,K}$ is

$$\hat{\Psi}_{KR}^{\alpha,K}(\omega) = 2\sinh(\alpha\sqrt{K^2 - \omega^2}) / (I_0(\alpha K)\sqrt{K^2 - \omega^2}),$$

if $\omega \in \mathbb{R} \setminus \{-K, K\}$ and $2\alpha/(I_0(\alpha K))$ otherwise.

The Kaiser Bessel window is a good and often used candidate for Ψ , since $\hat{\Psi}_{KB}^{\alpha,K}(\omega)$ becomes extremely small for $|\omega| \geq K$. For example, with the parameters K=3, and $\alpha=3\pi$, we have for $\omega \geq K$,

$$|\hat{\Psi}_{\rm KB}^{\alpha,K}(\omega)/\hat{\Psi}_{\rm KB}^{\alpha,K}(0)| \leq |\hat{\Psi}_{\rm KB}^{\alpha,K}(K)/\hat{\Psi}_{\rm KB}^{\alpha,K}(0)| \simeq 3*10^{-11}\,.$$

Remark 3. Take c=1 and let Ψ be the characteristic function of the interval $[-\pi,\pi]$. Then $\hat{\Psi}(\omega)=2\pi\operatorname{sinc}(\pi\omega)$ and (6), (7) reduce to the sinc series

$$\sum_{n=0}^{N-1} e^{-i\omega n 2\pi/N} g_n = \sum_{j \in \mathbb{Z}} e^{-i\pi(\omega-j)} \operatorname{sinc}(\omega-j) \left(\sum_{n=0}^{N-1} g_n e^{-ijn 2\pi/N} \right),$$

which is a discretized version of Shannon's sampling formula [39, 40]

$$\hat{g}(\omega) = \sum_{j \in \mathbb{Z}} e^{-i\pi(\omega - j)} \operatorname{sinc}(\omega - j) \hat{g}(j), \qquad \omega \in \mathbb{R},$$

applied to the Fourier transform of a function $g : \mathbb{R} \to \mathbb{R}$ that vanishes outside $[0, 2\pi]$.

See Figure 2 for a comparison of sinc and $\hat{\Psi}_{KB}^{\alpha,K}$, with K=3 and $\alpha=3\pi$. One realizes that $\hat{\Psi}_{KB}^{\alpha,K}$ decays much faster than sinc and is therefore much better suited for truncated interpolation. In fact, $|\hat{\Psi}_{KB}^{\alpha,K}(\omega)/\hat{\Psi}_{KB}^{\alpha,K}(0)| < 3*10^{-11}$ for $\omega \geq 3$, whereas $|\operatorname{sinc}(\omega)| < 0.01$ only for $\omega \geq 100/\pi$.

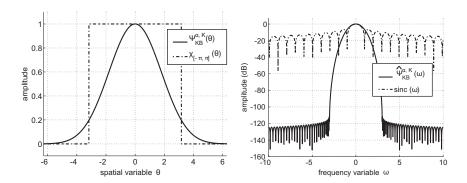


Figure 2: Left: Kaiser-Bessel window $\Psi_{\mathrm{KB}}^{\alpha,K}(\theta)$ and characteristic function of the interval $[-\pi,\pi]$. Right: Fourier transforms $\hat{\Psi}_{\mathrm{KB}}^{\alpha,K}(\omega)$ and $2\pi\operatorname{sinc}(\omega)$ in dB (decibel). Here dB denotes the logarithmic decay $10\log_{10}(|\phi(\omega)/\phi(0)|)$ of some quantity $\phi(\omega)$.

An error estimate for the nonuniform FFT using the Kaiser Bessel window is given in [29]. The result is

$$\left| e^{-i\omega\theta} - \frac{c}{2\pi\Psi(\theta)} \sum_{|\omega - j/c| < K} \hat{\Psi}_{KB}^{\alpha,K}(\omega - j/c) e^{-ij\theta/c} \right|$$

$$\leq \frac{30}{\pi I_0 (K\pi\sqrt{\alpha^2 - 1}/c^2)}.$$

For example, taking $c=2,~\alpha=3\pi$ and K=3, the above error is as small as $3*10^{-8}$.

4 A Fourier reconstruction Algorithm based on the nonuniform FFT

In this section we apply the one-dimensional nonuniform FFT to photoacoustic imaging. Throughout the following we restrict our attention to two dimensions, noting that the general case $d \geq 2$ can be treated in an analogous manner.

Assume that f is a smooth function that vanishes outside $(0, X)^2$, and set $g := w_{\text{cut}} \mathbf{Q} f$, where w_{cut} is as in (2). Fourier reconstruction names an implementation of (3), that uses discrete data

$$g_{m,n} := g(m\Delta_{\text{samp}}, n\Delta_{\text{samp}}),$$
 (9)

with $(m,n) \in \{0,\ldots,N-1\}^2$ and reconstructs an approximation

$$f_{m,n} \simeq f^{\dagger}(m\Delta_{\text{samp}}, n\Delta_{\text{samp}}),$$
 (10)

with $(m, n) \in \{0, ..., N-1\}^2$. Here f^{\dagger} is defined by (3), N is an even number and $\Delta_{\text{samp}} := X/N$. In the Appendix we show that the sampling in (10), (9) is

sufficiently fine, provided that $\Delta_{\text{samp}} \leq \pi/\Omega$, where Ω is the essential bandwidth of f.

Discretizing (3) with the trapezoidal rule gives

$$\sum_{n=0}^{N-1} \left(\sum_{m=0}^{N-1} e^{-i(\ln + km)2\pi/N} f_{m,n} \right)$$

$$= \frac{2k}{\omega_{k,l}} \sum_{n=0}^{N-1} e^{-i\omega_{k,l}n2\pi/N} \left(\sum_{m=0}^{N-1} e^{-ikm2\pi/N} g_{m,n} \right) , \quad (11)$$

where

$$\omega_{k,l} := \operatorname{sign}(l) \sqrt{k^2 + l^2}, \quad (k,l) \in \{-N/2, \dots, N/2 - 1\}^2.$$

One notices that the inner sums in (11),

$$\tilde{g}_{k,n} := \sum_{m=0}^{N-1} e^{-ikm2\pi/N} g_{m,n} \tag{12}$$

can be exactly evaluated with N one-dimensional FFTs, and the outer sums

$$\hat{g}_k(\omega_{k,l}) := \sum_{n=0}^{N-1} e^{-i\omega_{k,l} n 2\pi/N} \tilde{g}_{k,n}$$
(13)

can be approximately evaluated with N one-dimensional nonuniform FFTs. Denoting the resulting approximation by $\hat{g}_{k,l} \simeq \hat{g}_k(\omega_{k,l})$ and setting

$$\hat{f}_{k,l} := \frac{2k \, \hat{g}_{k,l}}{\omega_{k,l}}, \qquad (k,l) \in \{-N/2, \dots, N/2 - 1\}^2,$$
 (14)

we finally find

$$f_{n,m} := \frac{1}{N^2} \sum_{k,l=N/2}^{N/2-1} e^{i(km+ln)2\pi/N} \hat{f}_{k,l}$$
 (15)

with the inverse two-dimensional FFT algorithm.

The nonuniform FFT based reconstruction algorithm is summarized in Algorithm 2. Its numerical complexity can easily be estimated. Evaluating (12) requires $N\mathcal{O}(N\log N)$ operations (N one-dimensional FFTs), evaluating (13) requires $N\mathcal{O}(N\log N)$ operations (N non-uniform FFTs), and (15) is evaluated with the inverse two-dimensional FFT in $\mathcal{O}(N^2\log N)$ operations. Therefore the overall complexity of Algorithm 1 is $\mathcal{O}(N^2\log N)$.

In the next section we numerically compare Algorithm 2 with standard Fourier algorithms presented in the literature [41, 13], which all differ in the way how the sums in (13) are evaluated:

1. Direct Fourier algorithm. Equation (13) cannot be evaluated with the classical FFT algorithm because the nodes $\omega_{k,l}$ are non-equispaced.

Algorithm 2 Nonuniform FFT based algorithm for calculating $\mathbf{f} = (f_{m,n})_{n,m=0}^{N-1}$ using data $\mathbf{g} = (g_{m,n})_{m,n=0}^{N-1}$, oversampling factor c, interpolation length K, and window function Ψ .

```
1: \Psi \leftarrow \left(\Psi(2\pi n/N - \pi)\right)_n
2: \hat{\Psi} \leftarrow \left(e^{-i(\omega_k - j/c)\pi/c}\hat{\Psi}(\omega_k - j/c)\right)_{k,j}
                                                                                                                         > precomputations
  4: function FouRecNufft(\mathbf{g}, c, K, \mathbf{\Psi}, \hat{\mathbf{\Psi}})
              for n = 0, ..., N - 1 do
  5:
  6:
                    \mathbf{h} \leftarrow (g_{m,n})_m
                     (\tilde{g}_{k,n})_k \leftarrow \mathtt{fft}(\mathbf{h})
                                                                                                               ▷ one-dimensional FFT
  7:
              end for
  8:
              1 \leftarrow (-N/2, ..., N/2 - 1)
  9:
              for k = -N/2, ..., N/2 - 1 do
10:
                    \omega \leftarrow \text{sign}(\mathbf{l})\sqrt{k^2+\mathbf{l}^2}
11:
                    \mathbf{h} \leftarrow \text{nufft}(\mathbf{h}, \boldsymbol{\omega}, c, K, \boldsymbol{\Psi}, \hat{\boldsymbol{\Psi}})
                                                                                                                         ▷ nonuniform FFT
12:
                    (f_{k,l})_l \leftarrow 2k \, \mathbf{h}/\boldsymbol{\omega}
13:
              end for
14:
              \mathbf{f} \leftarrow (f_{k,l})_{k,l}
15:
              \mathbf{f} \leftarrow \mathtt{ifft2}(\mathbf{f})

    b two-dimensional inverse FFT

16:
              return f
17:
18: end function
```

The most simple way to evaluate (13) is with direct summation. Because there are N^2 such sums, direct Fourier reconstruction requires $\mathcal{O}(N^3)$ operations. Consequently it does not lead to a fast algorithm. However, since (13) is evaluated exactly, it is optimally suited to evaluate the image quality of reconstructions with fast Fourier algorithms.

2. Interpolation based algorithm. A fast and simple alternative to direct Fourier reconstruction is as follows: Choose an oversampling factor $c \ge 1$ and exactly evaluate

$$\hat{g}_k(\omega) := \Delta_{\text{samp}} \sum_{n=0}^{N-1} e^{-i\omega n 2\pi/N} \tilde{g}_{k,n} ,$$

at the uniformly spaced nodes $\omega = \Delta_{\text{samp}} j/c$, $j \in \{0, ..., Nc-1\}$, with the one-dimensional FFT algorithm. In a next step, linear interpolation is used to find approximate values $\hat{g}_{k,l} \simeq \hat{g}_k(\omega_{k,l})$, see [13]. Evaluating $\hat{g}_{k,l}$ with linear interpolation requires $\mathcal{O}(N^2)$ operation and therefore the overall numerical effort of linear interpolation based Fourier reconstruction is $\mathcal{O}(N^2 \log N)$.

Algorithms using nearest neighbor interpolation instead of the linear one have the same numerical complexity and have also been applied to PAI

(see, e.g., [42]). Higher order polynomial interpolation has been applied in [43] for a cubic recording geometry.

3. Truncated sinc reconstruction. If the function Ψ in Algorithm 2 is chosen as the characteristic function of the interval $[-c\pi, c\pi]$, $c \geq 1$, then the nonuniform fast Fourier transform reduces to the truncated sinc interpolation considered in [41]. However, due to the slow decay of $\operatorname{sinc}(\omega)$, truncation will introduce a non-negligible error in the reconstructed image (see Remark 3).

The Fourier algorithms are also be compared with a numerical implementation of the back-projection formula

$$f(x,y) = -\frac{2y}{\pi} \int_{\mathbb{R}} \left(\int_{r}^{\infty} \frac{(\partial_t t^{-1} \mathbf{Q} f)(x',t)}{\sqrt{t^2 - r^2}} dt \right) dx',$$
 (16)

where $r = \sqrt{(x-x')^2 + y^2}$ denotes the distance between the detector location (x',0) and the reconstruction point (x,y). Equation (16) has been obtained in [8] by applying the method of descent to the three-dimensional universal back-projection formula discovered by Xu and Wang [10]. Again (16) gives an exact reconstruction only if it is applied to complete data $(\mathbf{Q} f)(x,t)$, $(x,t) \in \mathbb{R}^2$. In the numerical experiments the back-projection formula is applied to the partial data $w_{\text{cut}} \mathbf{Q} f$, see (2), and implemented with $\mathcal{O}(N^3)$ operation counts as described in [8, Section 3.3].

5 Numerical Results

In the following we numerically compare the proposed nonuniform FFT based algorithm with standard Fourier algorithm and the back projection algorithm based on (16).

The cutoff function $w_{\rm cut}$ is constructed by convolution of

$$\varphi_{\epsilon}(x,t) = \begin{cases} C_{\epsilon} \exp(-1/(\epsilon - x^2 - t^2)^4), & \text{if } x^2 + t^2 < \epsilon, \\ 0, & \text{otherwise} \end{cases}$$

with the characteristic function of $[0,X]^2$, where ϵ is a small parameter and C_{ϵ} is chosen in such a way that $\int_{\mathbb{R}^2} \varphi_{\epsilon}(x,t) dx dt = 1$. Typically, ϵ is chosen as a "small" multiple of the sampling step size $\Delta_{\text{samp}} = X/N$.

In all numerical experiments we take X=1, and N=512. The window width α is chosen to be slightly smaller than $\pi(2c-1)$, where c is the oversampling factor that determines the accuracy of the Fourier reconstruction algorithms.

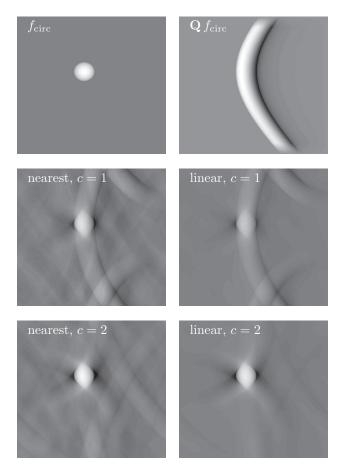


Figure 3: Reconstruction with interpolation based Fourier algorithms. White corresponds to function value 1, black to function value -0.4. Top Line: Phantom and analytic data. Middle line: Reconstruction without oversampling (c=1). Bottom line: Reconstruction with oversampling (c=2).

5.1 Circular shaped object

As first case example we use a circular shaped object

$$f_{\text{circ}}(\mathbf{x}) = \frac{2}{a} \left\{ \begin{array}{l} \left(a^2 - |\mathbf{x} - \mathbf{x}_0|^2\right)^{1/2}, & \text{if } |\mathbf{x} - \mathbf{x}_0| < a, \\ 0, & \text{otherwise,} \end{array} \right.$$

centered at $\mathbf{x}_0 := (x_0, y_0)$ (see top left image in Figure 3). For such a simple object reconstruction artifacts can be identified very clearly. Moreover, the data $\mathbf{Q} f_{\text{circ}}$ can be evaluated analytically (see [8, Equation (B.1)]) as

$$(\mathbf{Q} f_{\text{circ}})(x,t) = \frac{1}{a} \operatorname{Re} \left[(s_{+} - s_{-}) - t \log \left(\frac{s_{+} + (t + a_{i})}{s_{-} + (t - a_{i})} \right) \right].$$

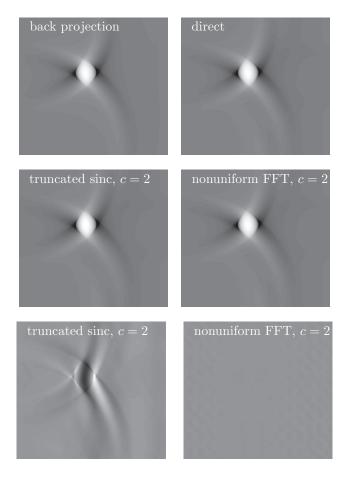


Figure 4: **Improved reconstructions.** Top Line: Back projection (left) and direct reconstruction (right). Middle Line: Truncated sinc (left) and nonuniform FFT based Fourier algorithm (right). Here white corresponds to function value 1, black to function value -0.4. Bottom Line: Difference images between direct and truncated sinc reconstruction (left), and direct and nonuniform FFT based reconstruction (right). Here white (resp. black) corresponds to function value 0.04 (resp. -0.04).

Here $s_{\pm} := ((t \pm a)^2 + |(x,0) - \mathbf{x}_0|^2)^{1/2}$, $\log(\cdot)$ is the principal branch of the complex logarithm, and Re[z] denotes the real part of complex number z. The reconstruction results are depicted in Figures 3 and 4. Table 1 and Figure 5 compare run times with the relative ℓ^2 -error

$$\frac{\|\mathbf{f} - \mathbf{f}^{\dagger}\|_{\ell^{2}}}{\|\mathbf{f}^{\dagger}\|_{\ell^{2}}} = \frac{\left(\sum_{m,n} (f_{m,n} - f_{m,n}^{\dagger})^{2}\right)^{1/2}}{\left(\sum_{m,n} (f_{m,n}^{\dagger})^{2}\right)^{1/2}},$$

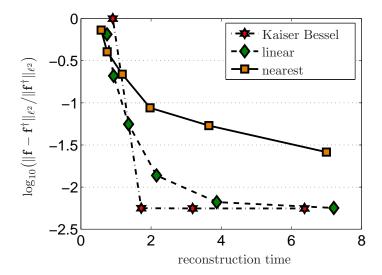


Figure 5: Reconstruction time versus error. The points on the graphs belong to runtimes and errors for reconstruction with oversampling factors $c \in \{1, 2, 4, 8, 16, 32\}$.

were $\mathbf{f}^{\dagger} = (f_{m,n}^{\dagger})$ denotes the discrete image obtained by direct Fourier reconstruction. Run times were measured for Matlab implementations on a personal computer with 2.4 GHz Athlon processor.

In order to demonstrate the stability of the Fourier algorithms, we also performed reconstructions from noisy data, where Gaussian noise was added with a variance equal to 20% of the maximal data value. The reconstruction results are depicted in Figure 7.

	c	ℓ^2 -error	runtime (sec)
back projection	-	-	88.9
direct reconstruction	-	-	54.1
nearest neighbor	1	0.75	0.65
nearest neighbor	2	0.40	0.85
linear	1	0.65	0.8
linear	2	0.21	0.95
Truncates sinc	2	0.04	1.6
Kaiser Bessel	2	0.006	1.6

Table 1: Run times and error of different reconstruction methods.

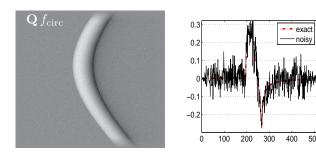


Figure 6: Noisy data used in Figure 7

5.2 Shepp-Logan phantom

In the next example we consider the Shepp-Logan phantom f_{phant} , which is shown in top left image in Figure 8. The data were calculated numerically by implementing d'Alemberts formula [44],

$$(\mathbf{Q} f_{\text{phant}})(x, 0, t) = \partial_t \int_0^t \frac{r(\mathbf{M} f_{\text{phant}})(x, 0, r)}{\sqrt{t^2 - r^2}} dr$$

with

$$(\mathbf{M} f_{\mathrm{phant}})(x, 0, r) := \frac{1}{2\pi} \oint_{|\sigma|=1} f_{\mathrm{phant}} ((x, 0) + r\sigma) d\sigma$$

denoting the spherical mean Radon transform of $f_{\rm circ}$. The reconstruction results from simulated data are depicted in Figure 8.

5.3 Discussion

We emphasize that none of the above Fourier algorithms are designed to calculate an approximation of f but an approximation to the partial Fourier reconstruction f^{\dagger} defined in (3). Therefore even in the direct reconstruction (top right image in Figure 4) and in the back projection reconstruction one can see some blurred boundaries in the reconstruction. Such artifacts are expected using limited view data (2), see [38, 37].

The results of interpolation based reconstruction without oversampling (c=1) are quite useless. The reconstructions are significantly improved by using a larger oversampling factor c. However, even then, the results never reach the quality of the nonuniform FFT based reconstruction. Moreover, the numerical effort of linear interpolation based reconstruction is proportional to the oversampling factor, which prohibits the use of "very large" values for c (see Figure 5). In the reconstruction with c=2 (bottom line in Figure 3 and middle line in Figure 8) artifacts are still clearly visible.

The images in the middle line of Figure 4 suggest that truncated sinc and nonuniform FFT based reconstruction seem to perform quite similar. However,

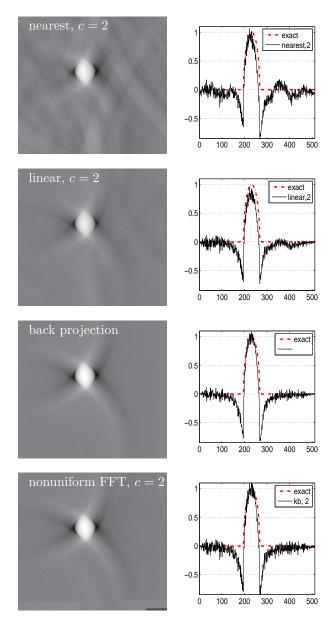


Figure 7: Reconstruction from noisy data. Top line: Nearest neighbor interpolation based reconstruction (c=2). Second line: Linear interpolation based reconstruction (c=2). Third line: Reconstruction with back projection algorithm. Bottom line: Reconstruction with nonuniform FFT algorithm (c=2). The horizontal profiles on the right are taken at $\{x=x_0\}$.

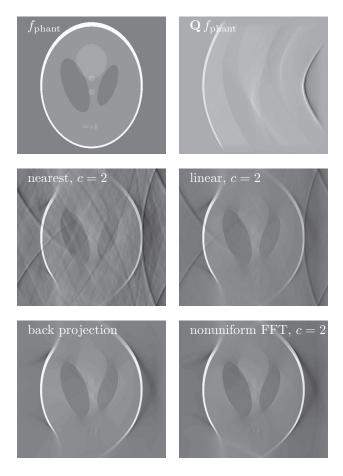


Figure 8: Reconstruction of Shepp Logan phantom. *Top line:* Phantom and simulated data. *Second line:* Interpolation based reconstruction. *Bottom line:* Reconstruction with back projection algorithm (left) and proposed nonuniform FFT algorithm (right).

the differences to the direct Fourier reconstructions, shown in the bottom line in Figure 4, demonstrate the higher accuracy of the nonuniform FFT based algorithm.

The results in Figure 7 show that all applied reconstruction algorithms are quite stable with respect to data perturbation. In particular, the filtered back projection algorithm produces images with the highest signal to noise ratio. However, only at the cost of a nearly 100 times longer computation time (see Table 1).

6 Conclusion

We presented a novel fast Fourier reconstruction algorithm for photoacoustic imaging using a limited planar detector array. The proposed algorithm is based on the nonuniform FFT. Theoretical investigation as well as numerical simulations show that our algorithm produces better images than existing Fourier algorithms with a similar numerical complexity. Moreover the proposed algorithm has been shown to be stable against data perturbations.

[Sampling and Resolution]

Let f be smooth function that vanishes outside $(0, X)^d$, and define g, f^{\dagger} by (2), (3). We further assume that $\mathbf{F} w_{\text{cut}}$ is concentrated around zero and, that f is essentially bandlimited with essential bandwidth Ω , in the sense that $(\mathbf{F} f)(\mathbf{K})$ is negligible for $|\mathbf{K}| \geq \Omega$. Note that since f has bounded support, $\mathbf{F} f$ cannot vanish exactly on $\{|\mathbf{K}| \geq \Omega\}$.

• Sampling of g. Equation (1) implies that

$$(\mathbf{F}g)(K_x,\omega) = (\mathbf{F}w_{\text{cut}}) * (\mathbf{F}\mathbf{Q}f)(K_x,\omega), \qquad (17)$$

with

$$(\mathbf{F} \mathbf{Q} f)(K_x, \omega) = \frac{2\omega(\mathbf{F} f) \left(K_x, \operatorname{sign}(\omega) \sqrt{\omega^2 - |K_x|^2}\right)}{\operatorname{sign}(\omega) \sqrt{\omega^2 - |K_x|^2}}$$

if $|\omega| > |K_x|^2$ and $(\mathbf{F} \mathbf{Q} f)(K_x, \omega) = 0$ otherwise. The assumption that f has essential bandwidth Ω and equation (17) imply that $(\mathbf{F} g)(K_x, \omega)$ is negligible outside the set

$$\{(K_x,\omega): |K_x| \leq |\omega| \leq \Omega\} \subset (-\Omega,\Omega)^d.$$

Now Shannon's sampling theorem [39, 40] states that g is sufficiently fine sampled if the step size in x and in t satisfies the Nyquist condition $\Delta_{\text{samp}} \leq \pi/\Omega$.

• Sampling of f^{\dagger} . Similar considerations as above again show that f^{\dagger} is essentially bandlimited with essential bandwidth Ω . Shannon's sampling theorem implies that f^{\dagger} can be reliable reconstructed from discrete samples taken with step size $\Delta_{\text{samp}} \leq \pi/\Omega$.

If f has essential bandwidth larger than Ω , the function g has to be filtered with a low pass-filter *before* sampling. Otherwise, sampling introduces aliasing artifacts in the reconstructed image.

Theoretically, the resolution (at least of the visible parts) can be increased ad infinity by simply decreasing the sampling size Δ_{samp} . In practical applications, several other factors such as the bandwidth of the ultrasound detection system limit the bandwidth of the data, and therefore the resolution of reconstructed images [33]. This, however, also guarantees that in practice a moderate sampling step size Δ_{samp} gives correct sampling without aliasing.

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